

## Article

# Characteristics of Offset Corbino Thin Film Transistor: A Physical Model

Jiaquan Kong, Chuan Liu, Xiaojie Li, Hai Ou, Juncong She, Shaozhi Deng  and Jun Chen \* 

State Key Laboratory of Optoelectronic Materials and Technologies, Guangdong Province Key Laboratory of Display Material and Technology, School of Electronics and Information Technology, Sun Yat-sen University, Guangzhou 510275, China; kongjq3@mail2.sysu.edu.cn (J.K.); liuchuan5@mail.sysu.edu.cn (C.L.); lixj67@mail2.sysu.edu.cn (X.L.); ohai@mail.sysu.edu.cn (H.O.); shejc@mail.sysu.edu.cn (J.S.); stdsdz@mail.sysu.edu.cn (S.D.)

\* Correspondence: stscjun@mail.sysu.edu.cn

**Abstract:** Offset Corbino thin film transistor is a good candidate for high voltage thin film transistor (HVTFT) due to the uniform drain electric field distribution benefiting from the circular structure. The physical model of offset Corbino thin film transistor characteristics has yet to be clarified. In this study, Equations are derived to describe the current–voltage relations of Corbino TFT with offset at the drain or source sides. The influence of offset position and parameters on the saturation voltage and the saturation current was described quantitatively. Three-dimensional Computer-Aided Design simulation and experiment results verify the theoretical physical model. Our physical model provides design rules for high voltage offset Corbino TFT when considering the voltage tolerance and saturation current balance.

**Keywords:** thin film transistor; Corbino; drain offset; high voltage; model



**Citation:** Kong, J.; Liu, C.; Li, X.; Ou, H.; She, J.; Deng, S.; Chen, J. Characteristics of Offset Corbino Thin Film Transistor: A Physical Model. *Electronics* **2023**, *12*, 2195. <https://doi.org/10.3390/electronics12102195>

Academic Editors: Martino Aldrigo, Ana-Maria Lepadatu, Florin Nastase and Andrei Avram

Received: 8 April 2023

Revised: 7 May 2023

Accepted: 9 May 2023

Published: 11 May 2023



**Copyright:** © 2023 by the authors. Licensee MDPI, Basel, Switzerland. This article is an open access article distributed under the terms and conditions of the Creative Commons Attribution (CC BY) license (<https://creativecommons.org/licenses/by/4.0/>).

## 1. Introduction

High voltage thin film transistors (HVTFT) have applications as driving components for field-emission display [1], dielectric elastomer actuators (DEA) [2], piezoelectric actuators [3], triboelectric nanogenerators [4], etc. Two approaches have been adopted to increase the breakdown voltage ( $V_{BD}$ ) of TFT. One is to use materials of wide bandgap ( $E_g > 3$  eV) as the active layer, such as ZnO, IGZO and  $Ga_2O_3$  [5,6]. The other is to use specific structures such as drain offset [7–9] or field-plate (FP) structure [10,11]. The drain offset structure uses the ungated channel region to undertake much of the drain voltage, which reduces the voltage on the gated channel region [12,13]. While in the field-plate structure, a plate located above the gate edge near the drain offset region is used to suppress the current collapse phenomena by restraining the gate edge electric field concentration [14–18]. Sometimes, these two approaches are used together to increase the breakdown voltage of TFT [19,20].

Many researchers have reported HVTFT based on conventional TFT with rectangular channels [21–26]. Martin, et al., fabricated high-voltage amorphous silicon TFT with a drain offset structure, operating at over 400 V [22]. Unagami, et al., reported high-voltage TFT fabricated in recrystallized polycrystalline silicon with a breakdown voltage above 100 V at an offset drain length of 20  $\mu\text{m}$  [23]. Yu, et al., used amorphous InGaZnO (a-IGZO) as the active layer and high-k  $Al_2O_3$  as the gate dielectric layer to fabricate HVTFT, which demonstrated a gate breakdown voltage ( $V_{BD,G}$ ) of 45–50 V and a source-drain channel breakdown voltage ( $V_{BD,SD}$ ) exceeding 100 V [24]. Chow, et al., fabricated rectangular a-Si HVTFT to operate voltage above 300 V with an offset of 6  $\mu\text{m}$  [25]. Li, et al., reported a-IGZO HVTFT with a high breakdown voltage of over 1.1 kV [26] and over 1.9 kV [27]. The results in the literature show that breakdown voltage can be improved if the drain offset structure is used. Although the drain offset structure can improve the breakdown voltage, it may reduce the operating current and occupy a larger area. The saturation

currents of rectangular TFTs with offset at the drain side are approximately the same, and at the source side decrease with offset length [12,21].

Corbino TFT is a circular TFT consisting of inner and outer concentric ring electrodes, superior to the rectangular-shaped TFT in some properties. Due to the uniform electric field distribution in the channel region, Corbino TFT can potentially work at higher drain bias [28,29] and achieve better mechanical bending stability [30,31]. In addition, the output resistance of Corbino TFT behaves almost infinitely beyond pinch-off with the outer-ring electrode as drain [32–34]. Mativenga, et al., showed that Corbino TFT could provide bending direction independence and achieve better stability under mechanical bending strain than rectangular-shaped TFTs [30]. Huo, et al., reported the fabrication of flexible and transparent IGZO Corbino TFT on muscovite mica substrates, which can keep electrical performances stable when exposed to tensile bending cycles up to 10,000 times [31]. Mativenga, et al., first reported a-IGZO Corbino TFT with almost infinite output resistance beyond pinch-off. They attributed this advantage to the nearly unchanged channel width to channel length ratio in Corbino TFT when the outer-ring electrode is the drain electrode [32]. Due to the high output resistance, Joo, et al., fabricated the high-gain complementary logic inverter composed of Corbino p-type SnO and n-type IGZO TFTs [33], and Geng, et al., reported a high-performance active image sensor pixel design by utilizing a-IGZO Corbino TFTs [34]. Furthermore, the concentric electrodes of Corbino TFT can be easily achieved using inkjet printing, taking advantage of the coffee ring effect [35,36].

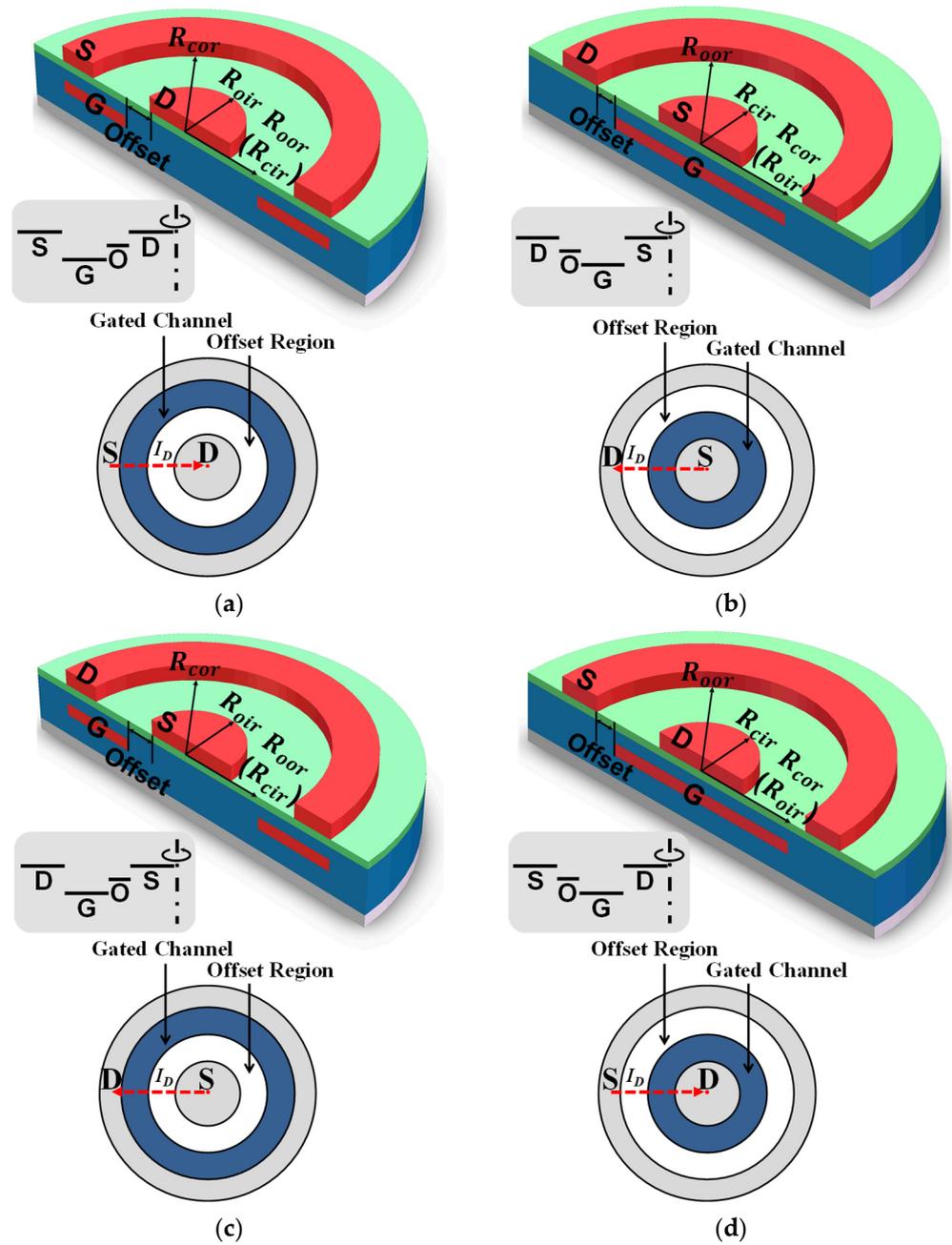
In contrast, research about high-voltage Corbino thin film transistors is limited [28,29]. Hong, et al., demonstrated magnesium zinc oxide (MZO) based Corbino HVTFT [28] and ZnO-based flexible Corbino HVTFT [29], which show a blocking voltage of 609 V and 150 V, respectively. As for the physical model of offset Corbino thin film transistor, Huo, et al., [19] gave a simplified expression of saturation voltage and saturation current for Corbino TFT with offset at the drain side. In contrast, the detailed theoretical analysis was not given, and the model of Corbino TFT with offset at the source side was not mentioned. The physical model of offset Corbino thin film transistor characteristics has not been clarified before. Obviously, the length and the position of offset play an essential role in the transfer characteristic and output characteristic of Corbino TFT, but the exact mechanism of how to offset influences the electrical characteristics, including the saturation current of Corbino TFT, has yet to be reported.

In this study, we put forward a physical model of offset Corbino TFT according to the basic resistance formula and electric potential analysis. The mathematical expression of current-voltage relations in offset Corbino TFT was derived. Both simulation and experiment are carried out to verify the model.

## 2. Device Structure of Model

According to the position of the offset region to the drain electrode, offset Corbino TFT can be classified into Drain-Offset Corbino TFT and Source-Offset Corbino TFT, in which the offset locates at the drain side and source side, respectively. The electrical characteristics of Corbino TFT vary when choosing the inner or outer ring electrode as the drain. Therefore, Drain-Offset Corbino TFT can be further divided into Source-Gate-Offset-Drain TFT (SGOD TFT) and Drain-Offset-Gate-Source TFT (DOGS TFT). Source-Offset Corbino TFT can be further divided into Drain-Gate-Offset-Source TFT (DGOS TFT) and Source-Offset-Gate-Drain TFT (SOGD TFT). The three-dimensional cross-sectional structures and top views of the four types of offset Corbino TFT are shown in Figure 1a–d. The inner and outer radiuses of the gated channel are denoted as  $R_{cir}$  and  $R_{cor}$ , and the inner and outer radiuses of the offset region are denoted as  $R_{oir}$  and  $R_{oor}$ . Due to the coupling of the gated channel region and offset region in offset Corbino TFT,  $R_{cir} = R_{oor}$  can be found in Figure 1a,c, and  $R_{oir} = R_{cor}$  can be found in Figure 1b,d. In addition, we label the length of the gated channel region and offset region as  $L_{channel}$  ( $R_{cor} - R_{cir}$ ) and  $L_{offset}$  ( $R_{oor} - R_{oir}$ ). Although SGOD TFTs and DOGS TFTs (also for DGOS TFTs and SOGD TFTs) have the same  $L_{channel}$  and  $L_{offset}$ , their gated channel and offset region differ due to the rotation symmetry of Corbino TFTs,

as shown in the top views. In addition, the direction of the drain current is also different for SGOD TFTs and DOGS TFTs. The charge carriers are injected from the outer electrode to the inner electrode in SGOD TFTs but from the inner electrode to the outer electrode in DOGS TFTs.



**Figure 1.** The three-dimensional cross-sectional structures and top views of offset Corbino TFTs. (a) SGOD TFT. (b) DOGS TFT. (c) DGOS TFT. (d) SOGD TFT. Insets of (a–d) are the corresponding schematic of offset Corbino TFTs.

### 3. Results and Discussion

#### 3.1. Formula

According to the gradual channel approximation, we can derive Equation (1) to describe the drain current of TFTs [37]:

$$I_D = \begin{cases} \frac{W}{L} \mu_1 C_i \left( V_G - V_{th} - \frac{V_D}{2} \right) V_D; \\ \frac{W}{L} \mu_1 C_i \left( V_G - V_{th} - \frac{V_1 + V_S}{2} \right) (V_1 - V_S). \end{cases} \quad (1)$$

The first form of Equation (1) describes the drain current of TFTs without offset, and the second form of Equation (1) is the general Equation to describe the current of gated channel applicable for all offset TFTs (rectangular and Corbino structure). Here,  $I_D$  is the drain current,  $W$  and  $L$  refer to the width and length of the gated semiconductor,  $\mu_1$  is the carrier mobility in the gated semiconductors,  $C_i = \epsilon_{ox}/t_{ox}$  with  $\epsilon_{ox}$  as permittivity and  $t_{ox}$  as the thickness of the insulating layer below or above the gated semiconductor, and  $V_G$  and  $V_{th}$  refer to the gate voltage and threshold voltage. The threshold voltage is the gate voltage axis intercept of the linear extrapolation of the transfer characteristics at its maximum first derivative (slope) point [38].  $V_D$  is the drain voltage,  $V_1$  is the potential at the end of gated semiconductors, and  $V_S$  is the source potential as transistors with organic or low-dimensional semiconductors usually have injection barriers, unlike MOSFETs. Therefore, the values of  $V_1$  and  $V_S$  depend on the specific structure of FETs with non-trivial gates. In addition, we derived the following Equation to describe the current of the offset region, calculated as the area of non-gated semiconductor times the current density [37]:  $I_D = Q_0 S \mu_2 (V_D - V_0 - V_1)^\alpha / d^\beta$ . Here,  $Q_0$  is the charge density factor, and  $\mu_2$  is the carrier mobility in the non-gated semiconductors.  $S$  and  $d$  are the current area and length of the non-gated semiconductor.  $V_0$  is the onset voltage for the non-gated channel (due to injection barriers or trap states). The injection barriers or trap states are related to the material of the active layer and metal electrode. Therefore,  $V_0$  is mainly related to material properties and film quality rather than transistor bias. Here,  $V_0$  is omitted for simplicity and could be included by replacing  $V_D$  with  $V_D - V_0$  whenever needed.  $\alpha$  is referred to as the charge transport factor and  $\beta \approx 2\alpha - 1$ . The current density of the offset region is the product of  $Q_0$  and  $\mu_2 (V_D - V_0 - V_1)^\alpha / d^\beta$ . Reference [35] shows our detailed analysis of  $Q_0$ ,  $\alpha$  and  $\beta$  according to different conduction mechanisms, including Ohmic, SCLC, and other cases. According to the continuity principle, the current of the gated channel is equal to that of the offset region, so we have

$$I_D = \frac{W}{L} \mu_1 C_i \left( V_G - V_{th} - \frac{V_1 + V_S}{2} \right) (V_1 - V_S) = Q_0 S \mu_2 \frac{(V_D - V_0 - V_1)^\alpha}{d^\beta}. \quad (2)$$

Here we consider the non-gated semiconductor with low resistance in the offset region. Thus, we assume that the carriers in the non-gated semiconductor of offset Corbino TFT obey Ohm's law. In this case, the value of  $\alpha$  is 1,  $\beta \approx 2\alpha - 1 = 1$ , and  $Q_0 = n_0 q$  with  $n_0$  as the intrinsic carrier concentration and  $q$  as the elementary charge. From this perspective, we can derive the physical model of Drain-Offset Corbino TFT and Source-Offset Corbino TFT by analyzing  $V_S$  and  $V_1$  in the corresponding situation.

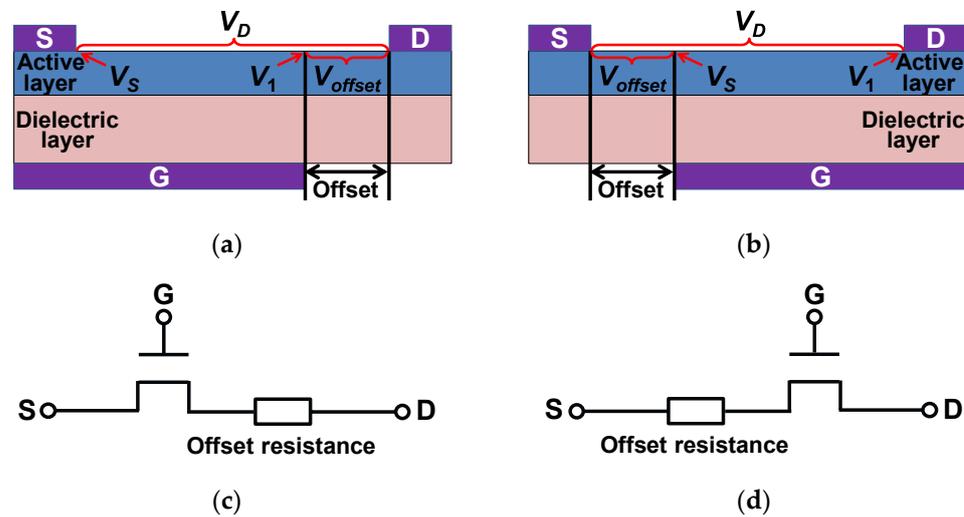
The cross-section image of Drain-Offset Corbino TFT and Source-Offset Corbino TFT are shown in Figure 2a,b. For either cross-section image structures, we can choose the drain pole or the source pole to rotate the circle's center to obtain two types of offset Corbino TFT.

Drain-Offset Corbino TFT could be equivalent to the series of traditional TFT and offset resistance, as shown in Figure 2c. We define  $R_{offset}$  as the non-gated offset region resistance, and the source serves as the reference point of the potential and is grounded instead of biased. We assume the contact in thin film-based transistors between the electrode and active layer is ohmic contact, and thus  $V_S$  is set as zero. Thus, we approximate  $V_1 + V_S$  and  $V_1 - V_S$  to  $V_1$  and obtain  $V_1 = V_D - I_D R_{offset}$ . By considering a radial electric field in Corbino TFT, the width-to-length ratio of Corbino TFT without offset can be written as

$2\pi / \ln(R_2/R_1)$  according to previous work [39], where  $R_2$  and  $R_1$  refer to the radius of the outer electrode and inner electrode. For Corbino offset resistance, we have:

$$R_{offset} = \frac{L_{offset}}{\sigma S} = \frac{L_{offset}}{q\mu_2 n_0 W_{offset} t} = \frac{\ln(R_{oor}/R_{oir})}{2\pi q\mu_2 n_0 t}. \quad (3)$$

Here,  $\sigma$  is electrical conductivity, and  $W_{offset}$  and  $t$  are the width and thickness of the offset semiconductor. It is worth noting that there are no geometric concepts of length and width in Corbino TFT with offset structure. Equation (3) considers that the geometrical factor of rectangular resistance is  $L_{offset}/W_{offset}$ , while Corbino-type offset resistance is  $\ln(R_{oor}/R_{oir})/2\pi$ . We know the gate voltage,  $V_G$ , could affect the carrier of the gated channel and the offset region. However, the Debye length in various semiconductors with low resistance at room temperature is in tens of nanometers. As a result, it is reasonable to ignore the effect of gate voltage when considering the resistance of the offset region of the micron scale.



**Figure 2.** The cross-section image of (a) Drain-Offset Corbino TFT, and (b) Source-Offset Corbino TFT. The schematic diagram of the equivalent circuit of (c) Drain-Offset Corbino TFT, and (d) Source-Offset Corbino TFT.

According to Equations (2) and (3), we have the drain current of Drain-Offset Corbino TFT in the linear region as:

$$I_D = K_n \left[ 2V_{GT} (V_D - I_D R_{offset}) - (V_D - I_D R_{offset})^2 \right]. \quad (4)$$

where  $V_{GT} = V_G - V_{th}$  and  $K_n$  refers to the transconductance coefficient of the device as  $\pi\mu_1 C_i / \ln(R_{cor}/R_{cir})$ . Equation (4) is rewritten into:

$$I_D = \left( -b + \sqrt{b^2 - 4ac} \right) / (2a) = \frac{V_D - V_{GT}}{R_{offset}} + \frac{\sqrt{\left( 2K_n R_{offset} V_{GT} + 1 \right)^2 - 4K_n R_{offset} V_D - 1}}{2K_n R_{offset}^2}. \quad (5)$$

where  $a = R_{offset}^2$ ,  $b = K_n^{-1} + 2V_{GT}R_{offset} - 2V_D R_{offset}$  and  $c = V_D^2 - 2V_{GT}V_D$ . From Equation (5), we can obtain the saturation drain current ( $I_{Dsat}$ ) and saturation drain voltage ( $V_{Dsat}$ ) when  $\partial I_D / \partial V_D = 0$ .

For TFTs in the saturation region, we denote the depletion length due to the channel length modulation effect as  $\Delta L$ . The  $\Delta L$  in a non-doped TFT is predicted to increase with  $V_D$  [37], similar to the classic doped MOSFETs [40]. For simplicity, we use a simple one-dimensional depletion model in doped MOSFET, i.e.,  $\Delta L \cong \sqrt{2\epsilon(V_D - V_{Dsat})/qN^*}$ . Here  $\epsilon$

is the permittivity of the active layer, and  $N^*$  is the effective dopant concentration. Unlike rectangular TFT, the position of the depletion region is significant for Corbino TFT, which is with drain near the inner or outer electrode. Therefore, we have the drain current in the saturation region in Corbino TFT with the drain at the inner electrode or outer electrode, respectively:

$$\frac{I_D}{I_{Dsat}} = \begin{cases} \frac{\ln(R_{cor}/R_{cir})}{\ln(R_{cor}/(R_{cir} + \Delta L))}, & \text{For SGOD and SOGD TFT;} \\ \frac{\ln(R_{cor}/R_{cir})}{\ln((R_{cor} - \Delta L)/R_{cir})}, & \text{For DOGS and DGOS TFT.} \end{cases} \quad (6)$$

Similarly, Source-Offset Corbino TFT could be equivalent to the series of traditional TFT and offset resistance, as shown in Figure 2d. However, in the case of Source-Offset Corbino TFT,  $V_1$  is  $V_D$ , and  $V_S$  is  $I_D R_{offset}$  in Equation (2). Therefore, we have the current-voltage ( $I$ - $V$ ) Equation of Source-Offset Corbino TFT in the linear region as:

$$I_D = K_n \left( 2V_{GT} - V_D - I_D R_{offset} \right) \left( V_D - I_D R_{offset} \right). \quad (7)$$

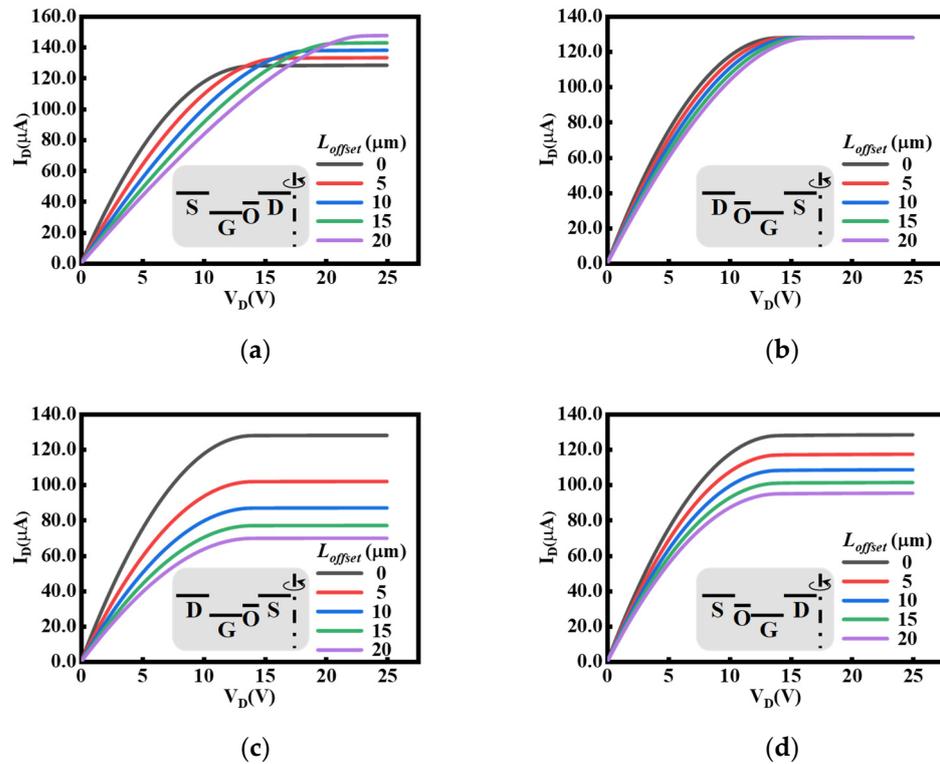
From Equations (6) and (7),  $I_{Dsat}$ ,  $V_{Dsat}$ , and the drain current of Source-Offset Corbino TFT in the saturation region could be obtained likewise.

We derive the current-voltage ( $I$ - $V$ ) Equations (listed in Table 1) of four types of offset Corbino TFT in the linear region and saturation region independent of the active layer material, from which we can derive important parameters such as the saturation drain voltage ( $V_{Dsat}$ ) and the saturation drain current ( $I_{Dsat}$ ). The position and length of the offset region are the key parameters influencing the electrical characteristics. Therefore, the derived equations listed in Table 1 are helpful when designing offset Corbino TFT for specific applications.

**Table 1.** The current-voltage equations of four types of offset Corbino TFT and other important parameters.

Drain-Offset Corbino TFT		
	SGOD TFT	DOGS TFT
$V_1$	$V_D - I_D R_{offset}$	
$V_S$	0	
$I_D$ in the linear region	$\frac{V_D - V_{GT}}{R_{offset}} + \frac{\sqrt{(2K_n R_{offset} V_{GT} + 1)^2 - 4K_n R_{offset} V_D} - 1}{2K_n R_{offset}^2}$	
$I_D$ in the saturation region	$I_{Dsat} \frac{\ln\left(\frac{R_{cor}}{R_{cir}}\right)}{\ln\left(\frac{R_{cor}}{R_{cir} + \Delta L}\right)}$	$I_{Dsat} \frac{\ln\left(\frac{R_{cor}}{R_{cir}}\right)}{\ln\left(\frac{R_{cor} - \Delta L}{R_{cir}}\right)}$
$V_{Dsat}$	$K_n R_{offset} V_{GT}^2 + V_{GT}$	
$I_{Dsat}$	$K_n V_{GT}^2$	
Source-Offset Corbino TFT		
	DGOS TFT	SOGD TFT
$V_1$	$V_D$	
$V_S$	$I_D R_{offset}$	
$I_D$ in the linear region	$\frac{V_{GT}}{R_{offset}} - \frac{\sqrt{4K_n^2 R_{offset}^2 (V_{GT} - V_D)^2 + 4K_n R_{offset} V_{GT} + 1} - 1}{2K_n R_{offset}^2}$	
$I_D$ in the saturation region	$I_{Dsat} \frac{\ln\left(\frac{R_{cor}}{R_{cir}}\right)}{\ln\left(\frac{R_{cor} - \Delta L}{R_{cir}}\right)}$	$I_{Dsat} \frac{\ln\left(\frac{R_{cor}}{R_{cir}}\right)}{\ln\left(\frac{R_{cor}}{R_{cir} + \Delta L}\right)}$
$V_{Dsat}$	$V_{GT}$	
$I_{Dsat}$	$\frac{V_{GT}}{R_{offset}} - \frac{\sqrt{4K_n R_{offset} V_{GT} + 1} - 1}{2K_n R_{offset}^2}$	

The theoretical output characteristics of 4 types of offset Corbino TFT were drawn for different  $L_{offset}$ , and the results are shown in Figure 3. The corresponding structure parameters and other parameters used are given in Table 2. For Drain-Offset Corbino TFTs, the  $V_{Dsat}$  increases with  $L_{offset}$ . The  $I_{Dsat}$  of DOGS TFT remains unchanged independent of  $L_{offset}$ . The  $I_{Dsat}$  of SGOD TFT is larger for longer  $L_{offset}$  because of larger  $K_n$  for longer  $L_{offset}$  in SGOD TFT with the same  $L_{channel}$ . For Source-Offset Corbino TFT, including DGOS TFT and SOGD TFT,  $I_{Dsat}$  decreases with longer  $L_{offset}$  while  $V_{Dsat}$  remains unchanged. As shown in Figure 3, the primary influence of offset in Drain-Offset Corbino TFT is to increase the saturation voltage, and in Source-Offset Corbino TFT is to reduce the saturation current. The offset in Drain-Offset Corbino TFT undertakes a great part of drain voltage, and thus Drain-Offset Corbino TFT can operate at higher drain voltage. The SGOD TFT can endure higher potential due to a larger  $R_{offset}$  compared with DOGS TFT with the same  $L_{offset}$  and  $L_{channel}$ . Our derived model can also demonstrate the influence of the inherent conductivity of semiconductors (such as  $n_0, \mu_1$  and  $\mu_2$ ) on  $I_D$ . When the semiconductor is more conductive, the  $I_D$  of offset Corbino TFT is higher. The influence of offset on drain current shown in transfer characteristics is the same as in output characteristics, so we only show the relevant output curves in the manuscript for a concise presentation.

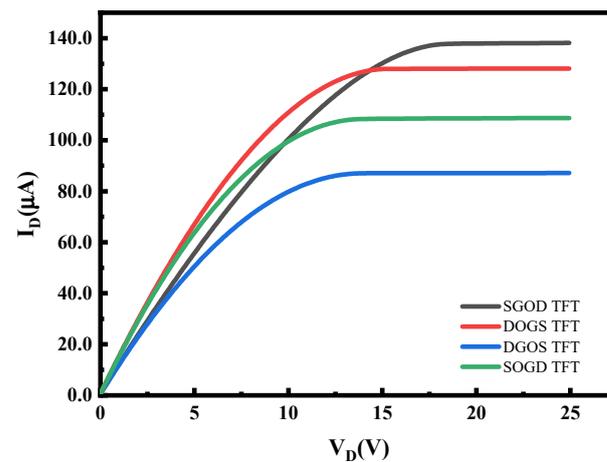


**Figure 3.** The theoretical output characteristics of offset Corbino TFT when  $V_G$  is 15 V: (a) SGOD TFT. (b) DOGS TFT. (c) DGOS TFT. (d) SOGD TFT. Insets of (a–d) are the corresponding schematic of offset Corbino TFT.

The position of the offset region affects the current–voltage characteristics of offset Corbino TFT, as shown in Figure 4. With the same  $L_{offset}$  and  $L_{channel}$ , the  $I_{Dsat}$  of Source-Offset Corbino TFT is smaller than that of Drain-Offset Corbino TFT because the offset in Source-Offset Corbino TFT reduces the saturation current. The DGOS TFT has the lowest  $I_{Dsat}$ , and the SOGD TFT has the second lowest  $I_{Dsat}$ . In the formula of  $I_{Dsat}$  of Source-Offset Corbino TFT, the value of  $R_{offset}$  exists as a denominator. The  $V_{Dsat}$  and  $I_{Dsat}$  of SGOD TFT are greater than those of DOGS TFT on account of larger  $K_n$  and  $R_{offset}$ , which can be explained directly using the formula of  $V_{Dsat}$  and  $I_{Dsat}$  of Drain-Offset Corbino TFT shown in Table 1.

**Table 2.** The structure parameters and other parameters of offset Corbino TFT used for theoretical characteristics plotting.

Parameter	Value	Parameter	Value
$V_{th}$ (V)	1.0	$n_0$ (cm <sup>-3</sup> )	$1.5 \times 10^{17}$
$\mu_1$ (cm <sup>2</sup> V <sup>-1</sup> s <sup>-1</sup> )	5.0	$\mu_2$ (cm <sup>2</sup> V <sup>-1</sup> s <sup>-1</sup> )	5.0
$C_i$ (F/cm <sup>2</sup> )	$5.0 \times 10^{-8}$	$t$ (cm)	$0.05 \times 10^{-4}$
$\epsilon$ (F/cm)	$8.5 \times 10^{-13}$	$N^*$ (cm <sup>-3</sup> )	$1.0 \times 10^{17}$
$L_{offset}$ (μm)	0, 5, 10, 15, 20	$L_{channel}$ (μm)	175
Inner electrode radius (μm)	75		

**Figure 4.** The theoretical output characteristics of different offset Corbino TFTs with the same  $L_{offset}$  of 10 μm and the same  $L_{channel}$  of 175 μm when  $V_G$  is 15 V.

It is worth mentioning that, as shown in Figures 3 and 4, ideal Corbino TFTs have almost infinite output resistance when the outer-ring electrode is the drain, which has been reported in previous literature [32]. As for this phenomenon, the expression of  $I_D$  in the saturation region shown in Table 1 gives direct quantitative theoretical proof. Channel length modulation in the device leads to variations in the  $I_D$  of offset Corbino TFT in the saturation region. As shown in Table 1, the  $I_D$  of offset Corbino TFT in the saturation region relates to the  $R_{cor}$  and  $R_{cir}$ . In our one-dimensional depletion model, the same drain voltage increment in the saturation region results in the same  $\Delta L$ . In the case of  $R_{cor}$  to  $R_{cir}$  ratio greater than one and with the same  $\Delta L$ , our Equations show that  $I_D$  of Corbino TFT with the drain at the outer electrode in the saturation region increases more slowly than that of Corbino TFT with the drain at the inner electrode. The output conductance is expressed as  $\partial I_D / \partial V_D$ , so the Corbino TFT with the drain at the outer electrode has higher output resistance than Corbino TFT with the drain at the inner electrode.

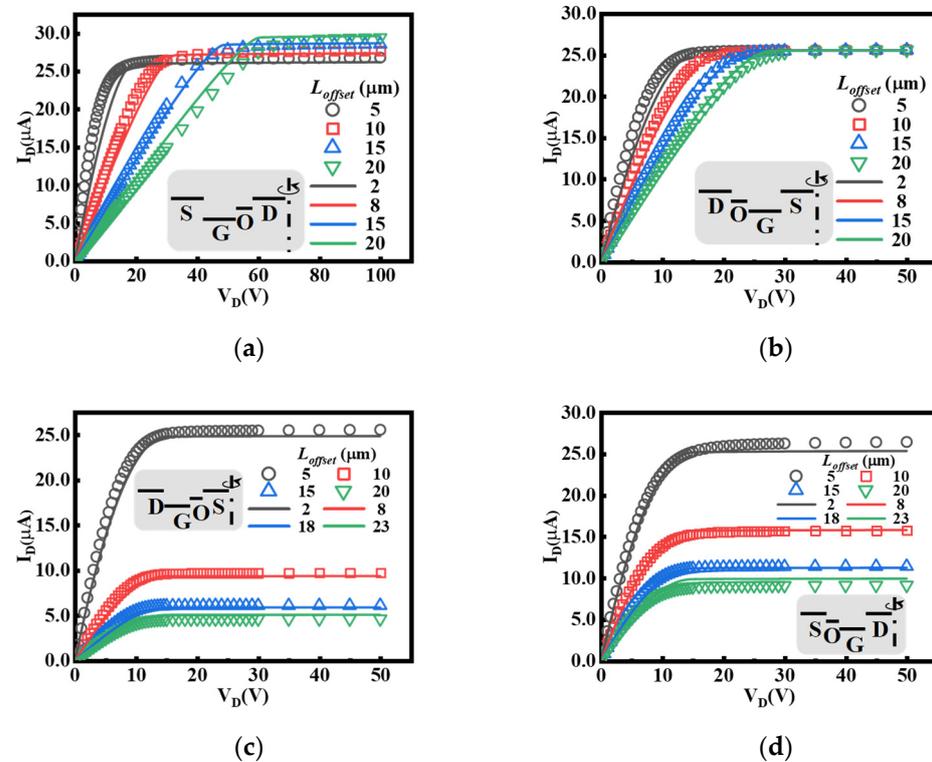
### 3.2. Simulation

In order to further validate the theoretical derivation, we used the three-dimensional technology Computer-Aided Design (TCAD) by Silvaco tool to simulate the characteristics of offset Corbino TFT. The material we choose as the active layer is amorphous indium-gallium-zinc-oxide (a-IGZO). For the TCAD simulation, density of states (DOS) parameters such as conduction and valence band tail states ( $N_{TD}$  and  $N_{TA}$ ), conduction and valence band tail state slopes ( $W_{TD}$  and  $W_{TA}$ ), donor-like and acceptor-like states ( $N_{GD}$  and  $N_{GA}$ ) with Gaussian distributions, and their respective full width at half maximums ( $W_{GD}$  and  $W_{GA}$ ) are listed in Table 3. The corresponding structure parameters for simulation are the same as the theoretical characteristics plotting, as shown in Table 2.

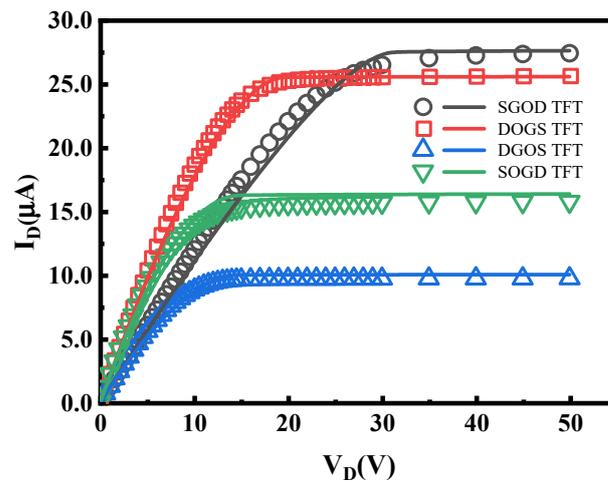
**Table 3.** Silvaco TCAD simulation density of states parameters for IGZO.

Parameter	Value	Parameter	Value
$N_{TA} (\text{cm}^{-3}\text{eV}^{-1})$	$5.5 \times 10^{18}$	$N_{GA} (\text{cm}^{-3}\text{eV}^{-1})$	$1.85 \times 10^{16}$
$W_{TA} (\text{eV}^{-1})$	0.015	$W_{GA} (\text{eV}^{-1})$	0.3
$N_{TD} (\text{cm}^{-3}\text{eV}^{-1})$	$5.5 \times 10^{18}$	$N_{GD} (\text{cm}^{-3}\text{eV}^{-1})$	$1.225 \times 10^{17}$
$W_{TD} (\text{eV}^{-1})$	0.22	$W_{GD} (\text{eV}^{-1})$	0.13

TCAD simulation results also show that offset has a similar effect on Corbino TFT with theoretical results. The simulated and theoretical fitting output characteristics of 4 types of offset Corbino TFT are shown in Figure 5. The theoretical fitting parameters used different from Table 2 are  $V_{th} = 0 \text{ V}$ ,  $C_i = 8.7 \times 10^{-9} \text{ F/cm}^2$ ,  $n_0 = 5.0 \times 10^{16} \text{ cm}^{-3}$  for the curves of  $L_{offset} = 2 \mu\text{m}$  in Figure 5c,d and  $n_0 = 3.8 \times 10^{15} \text{ cm}^{-3}$  for other fitting curves in Figure 5s and 6,  $\mu_2 = 11 \text{ cm}^2\text{V}^{-1}\text{s}^{-1}$  for Figure 6 and  $\mu_2 = 8 \text{ cm}^2\text{V}^{-1}\text{s}^{-1}$  for other fitting curves in Figure 5. Comparing Figure 5 with Figure 3, we found that the simulation results show that the characteristics of the TFT exhibit similar trends when offset length ( $L_{offset}$ ) increases. Offset in Drain-Offset Corbino TFT makes no difference to  $I_{Dsat}$  with the same  $K_n$ , as shown in Figure 5b.  $I_{Dsat}$  decreases with offset length more rapidly in Source-Offset Corbino TFT, as shown in Figure 5c,d. Longer  $L_{offset}$  significantly increases  $V_{Dsat}$  in Drain-Offset Corbino TFT, as shown in Figure 5a,b. In the same way, we study the effect of the position of the offset region with the same  $L_{offset}$  and  $L_{channel}$  by TCAD simulation, as shown in Figure 6. The feature and trend of  $V_{Dsat}$  and  $I_{Dsat}$  with offset in Figure 6 are consistent with Figure 4, although their exact values are different. We can conclude that the TCAD simulation results prove the correctness of the formula of offset Corbino TFT we derived.



**Figure 5.** The simulation output characteristics of offset Corbino TFT when  $V_G$  is 15 V: (a) SGOD TFT. (b) DOGS TFT. (c) DGOS TFT. (d) SOGD TFT. Insets of (a–d) are the corresponding schematic of offset Corbino TFT. The TCAD simulated data are drawn in dots, and relevant results using the derived Equations are drawn in curves.



**Figure 6.** The simulation output characteristics of different offset Corbino TFTs with the same  $L_{offset}$  of  $10\ \mu\text{m}$  and the same  $L_{channel}$  of  $175\ \mu\text{m}$  when  $V_G$  is  $15\ \text{V}$ . The TCAD simulated data are drawn in dots, and fitting results by using the derived Equations are drawn in curves.

Although TCAD simulation and theoretical formula help design and predict offset Corbino TFT characteristics, the physical model's theoretical formula is more convenient and practical. First, the three-dimensional Silvaco TCAD simulation of offset Corbino TFT is time-consuming. We could not simplify the three-dimensional Corbino TFT simulation to a two-dimensional one to save software run time because it is not translational symmetrical in Corbino TFT. Even if we can simplify to two-dimensional TCAD simulation, formula calculation without calculus operation always runs faster than TCAD simulation. Secondly, TCAD simulation may encounter problems such as limiting materials and non-convergence in operation. Thus, we could employ formulas to help design device characteristics in most cases.

### 3.3. Experimental Verification

Apart from theoretical Equation derivation and TCAD simulation, experiment results can also add credibility to the physical model of offset Corbino TFT. We fabricated a-IGZO offset Corbino TFT to verify the physical model. The fabricated a-IGZO Corbino TFTs have a bottom-gate, inverted-staggered structure prepared on glass substrates. The circular gate electrode was a  $200\ \text{nm}$  thick Mo layer by depositing and patterning via wet etching. The gate insulator of  $400\ \text{nm}$  thick  $\text{SiO}_2$  was deposited by plasma-enhanced chemical vapor deposition (PECVD). A  $50\text{-nm}$ -thick IGZO film was sputtered and etched to form the active layer on top of the gate insulator. Both the source and drain electrodes were formed by sputtering and lift-off processes. A  $\text{SiO}_2$  passivation layer was deposited by PECVD and etched by reactive ion etching (RIE) to form contact holes. We fill the contact holes with indium tin oxide (ITO) by magnetron sputtering and patterning it via a lift-off process. There are two annealing steps in device fabrication. IGZO film was annealed by thermal oxidation at  $350\ ^\circ\text{C}$  for 1 hour in  $\text{N}_2$  atmosphere, and ITO was annealed by thermal oxidation at  $470\ ^\circ\text{C}$  for 2 hours in air atmosphere.

We fabricate four types of offset Corbino TFT with 15 different values of  $L_{offset}$ , i.e.,  $0, 5, 10, 15, 20, 25, 30, 35, 40, 45, 50, 55, 60, 75, 100\ \mu\text{m}$ . The optical images of two fabricated devices are shown in Figure 7. The electrical characteristics of the a-IGZO Corbino TFT were measured using a semiconductor analyzer (Agilent B1500A). The typical experimental and theoretical fitting output characteristics of fabricated offset Corbino TFTs are shown in Figure 8. The theoretical fitting parameters different from Table 2 are  $V_{th} = -1\ \text{V}$ ,  $n_0 = 3.8 \times 10^{15}\ \text{cm}^{-3}$ ,  $\mu_2 = 2.7\ \text{cm}^2\text{V}^{-1}\text{s}^{-1}$ ,  $C_i = 6.4 \times 10^{-9}\ \text{F}/\text{cm}^2$ ,  $\mu_1 = 2.7\ \text{cm}^2\text{V}^{-1}\text{s}^{-1}$  for Figure 8a and  $\mu_1 = 4\ \text{cm}^2\text{V}^{-1}\text{s}^{-1}$  for Figures 8b–d and 9. Different positions of offset region in offset Corbino TFT with the same  $L_{offset}$  and  $L_{channel}$  are also studied in the experiment.

The experimental and theoretical fitting output characteristics of different types of offset Corbino TFT with the same  $L_{offset}$  and  $L_{channel}$  are shown in Figure 9.

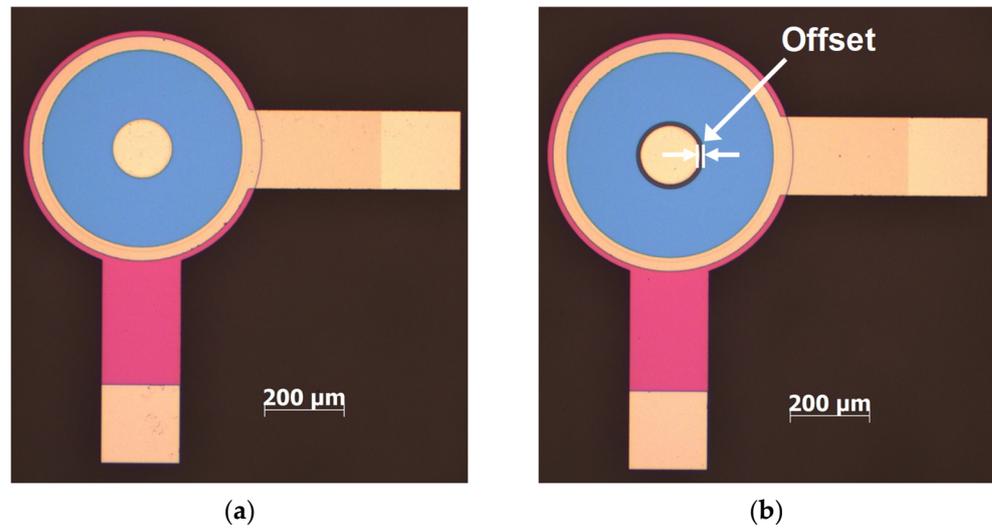


Figure 7. Optical images of fabricated a-IGZO Corbino TFTs: (a) without offset; and (b) with an offset of 10  $\mu\text{m}$ .

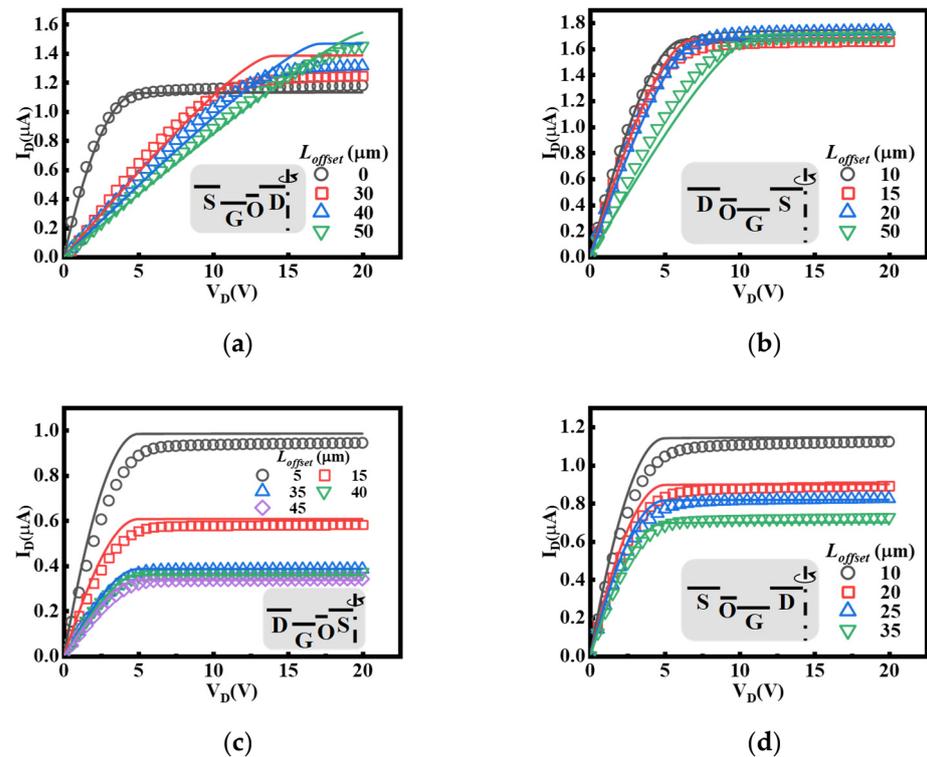
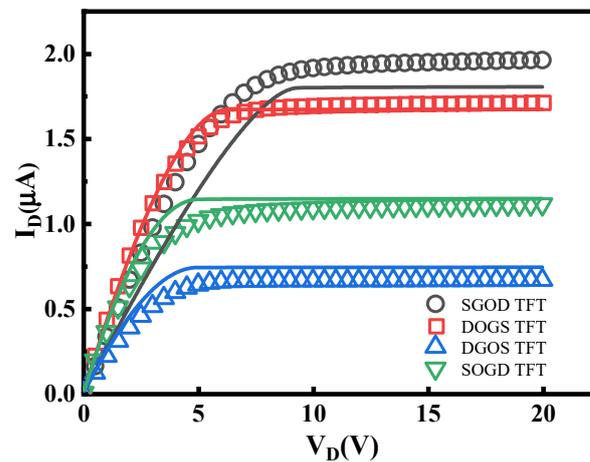


Figure 8. The experimental output characteristics of offset Corbino TFT when  $V_G$  is 4 V: (a) SGOD TFT. (b) DOGS TFT. (c) DGOS TFT. (d) SOGD TFT. Insets of (a–d) are the corresponding schematic of offset Corbino TFT. The dots are experimental data, and the curves are the fitting results using the derived Equations.

For Drain-Offset Corbino TFT, the experiment results in Figure 8a,b show that the  $V_{Dsat}$  increases with the  $L_{offset}$ . SGOD TFT with longer  $L_{offset}$  has larger  $I_{Dsat}$ , and different DOGS TFT almost have the same  $I_{Dsat}$  despite different  $L_{offset}$ . Despite the slight variance in  $I_{Dsat}$ , the experimental results verify the theoretical derivation and TCAD simulation, where

different DOGS TFT have identical  $I_{Dsat}$  for different values of  $L_{offset}$ . For Source-Offset Corbino TFT, experimental TFTs with longer  $L_{offset}$  have smaller  $I_{Dsat}$ . In addition, the  $V_{Dsat}$  of experimental Source-Offset Corbino TFTs is almost the same despite different  $L_{offset}$ , as demonstrated in the formula. In Figure 9, the regulation of  $V_{Dsat}$  and  $I_{Dsat}$  with offset is the same as expected from the theoretical formula and simulation. Overall, the experiment results indicate that our physical model of offset Corbino TFT is reasonable.



**Figure 9.** The experimental output characteristics of different offset Corbino TFTs with the same  $L_{offset}$  of 10  $\mu\text{m}$  and the same  $L_{channel}$  of 175  $\mu\text{m}$  when  $V_G$  is 4 V. The dots are experimental data, and the curves are the fitting results by using the derived Equations.

#### 4. Conclusions

According to the basic resistance formula and electric potential analysis, we derived a physical model of offset Corbino TFT, including Drain-Offset Corbino TFT and Source-Offset Corbino TFT. We explored how the offset's position and length influence the transfer and output characteristics of Corbino TFT. For Drain-Offset Corbino TFT, the saturation voltage increases with offset length, but the saturation current remains the same as long as the gated channel region is the same. When offset length increases, Source-Offset Corbino TFT has the same saturation voltage but a much smaller saturation current. TCAD simulation and experiment results verified the physical model of offset Corbino TFT. Our physical model is beneficial to device design for which the saturation voltage and saturation current must be considered. Offset Corbino TFT is a good candidate for high-voltage applications, and our model could provide guidelines for balancing voltage tolerance and the saturation current in device design.

**Author Contributions:** J.C. and C.L. conceived the idea and proposed the experimental scheme. J.C., C.L. and J.K. initiated this study. J.C., C.L. and J.K. established the model. J.K. carried out the simulations. J.K., X.L. and H.O. carried out the experiments. J.C., C.L., S.D., J.S. and J.K. discussed and interpreted the results. J.C., C.L. and J.K. co-wrote the manuscript. All authors have read and agreed to the published version of the manuscript.

**Funding:** This research was funded by the National Key Research and Development Program of China (Grant No. 2022YFA1204200), the National Natural Science Foundation of China (Grant No. 91833303 and 61922090), the Science and Technology Department of Guangdong Province (Grant No.2020B0101020002), the Natural Science Foundation of Guangdong Province (Grant No. 2018B030311045), Fundamental Research Funds for the Central Universities, and the Guangzhou Science Technology and Innovation Commission.

**Data Availability Statement:** The data that support the findings of this study are available from the corresponding author upon reasonable request.

**Conflicts of Interest:** The authors declare no conflict of interest.

## References

1. Song, Y.H.; Kim, K.B.; Hwang, C.S.; Park, D.J.; Lee, J.H.; Kang, K.Y.; Hur, J.H.; Jang, J. Active-matrix field-emission display based on a CNT emitter and a-Si TFTs. *J. Soc. Inf. Disp.* **2005**, *13*, 241–244. [[CrossRef](#)]
2. Marette, A.; Poulin, A.; Besse, N.; Rosset, S.; Briand, D.; Shea, H. Flexible zinc–tin oxide thin film transistors operating at 1 kV for integrated switching of dielectric elastomer actuators arrays. *Adv. Mater.* **2017**, *29*, 1700880. [[CrossRef](#)] [[PubMed](#)]
3. Karpelson, M.; Wei, G.-Y.; Wood, R.J. Driving high voltage piezoelectric actuators in microrobotic applications. *Sens. Actuators A* **2012**, *176*, 78–89. [[CrossRef](#)]
4. Zhang, Y.; Mei, Z.; Wang, T.; Huo, W.; Cui, S.; Liang, H.; Du, X. Flexible transparent high-voltage diodes for energy management in wearable electronics. *Nano Energy* **2017**, *40*, 289–299. [[CrossRef](#)]
5. Galazka, Z.; Uecker, R.; Irmscher, K.; Albrecht, M.; Klimm, D.; Pietsch, M.; Brützm, M.; Bertram, R.; Ganschow, S.; Fornari, R. Czochralski growth and characterization of  $\beta$ -Ga<sub>2</sub>O<sub>3</sub> single crystals. *Cryst. Res. Technol.* **2010**, *45*, 1229–1236. [[CrossRef](#)]
6. Hasegawa, H.; Kawabe, U.; Aita, T.; Ishiba, T. Single crystal growth of layered perovskite metal oxides. *Jpn. J. Appl. Phys.* **1987**, *26*, L673. [[CrossRef](#)]
7. Seki, S.; Kogure, O.; Tsujiyama, B. Leakage current characteristics of offset-gate-structure polycrystalline-silicon MOSFET's. *IEEE Electron Device Lett.* **1987**, *8*, 434–436. [[CrossRef](#)]
8. Unagami, T. High-voltage poly-Si TFTs with multichannel structure. *IEEE Trans. Electron Devices* **1988**, *35*, 2363–2367. [[CrossRef](#)]
9. Huang, T.-Y.; Lewis, A.; Wu, I.-W.; Chiang, A.; Bruce, R. New intra-gate-offset high-voltage thin-film transistor with misalignment immunity. *Electron. Lett.* **1989**, *25*, 544–545. [[CrossRef](#)]
10. Huang, T.-Y.; Wu, I.-W.; Lewis, A.G.; Chiang, A.; Bruce, R.H. A simpler 100-V polysilicon TFT with improved turn-on characteristics. *IEEE Electron Device Lett.* **1990**, *11*, 244–246. [[CrossRef](#)]
11. Huang, T.-Y.; Wu, I.-W.; Lewis, A.; Chiang, A.; Bruce, R. Device sensitivity of field-plated polysilicon high-voltage TFTs and their application to low-voltage operation. *IEEE Electron Device Lett.* **1990**, *11*, 541–543. [[CrossRef](#)]
12. Park, C.; Billah, M.M.; Siddik, A.B.; Lee, S.; Han, B.; Jang, J. High Voltage Amorphous InGaZnO TFT with F Doped Drain Offset Structure. *IEEE Electron Device Lett.* **2021**, *42*, 1476–1479. [[CrossRef](#)]
13. Wu, M.-H.; Lin, H.-C.; Li, P.-W. Film-profile-engineered ZnO thin-film transistor with gate/drain offset for high-voltage operation. *Jpn. J. Appl. Phys.* **2019**, *58*, 066502. [[CrossRef](#)]
14. Saito, W.; Kakiuchi, Y.; Nitta, T.; Saito, Y.; Noda, T.; Fujimoto, H.; Yoshioka, A.; Ohno, T.; Yamaguchi, M. Field-plate structure dependence of current collapse phenomena in high-voltage GaN-HEMTs. *IEEE Electron Device Lett.* **2010**, *31*, 659–661. [[CrossRef](#)]
15. Saito, W.; Nitta, T.; Kakiuchi, Y.; Saito, Y.; Tsuda, K.; Omura, I.; Yamaguchi, M. On-resistance modulation of high voltage GaN HEMT on sapphire substrate under high applied voltage. *IEEE Electron Device Lett.* **2007**, *28*, 676–678. [[CrossRef](#)]
16. Zhang, P.; Zhao, S.-L.; Hou, B.; Wang, C.; Zheng, X.-F.; Ma, X.-H.; Zhang, J.-C.; Hao, Y. Improvement of the off-state breakdown voltage with field plate and low-density drain in AlGaIn/GaN high-electron mobility transistors. *Chin. Phys. B* **2015**, *24*, 037304. [[CrossRef](#)]
17. Saito, W.; Nitta, T.; Kakiuchi, Y.; Saito, Y.; Tsuda, K.; Omura, I.; Yamaguchi, M. Suppression of dynamic on-resistance increase and gate charge measurements in high-voltage GaN-HEMTs with optimized field-plate structure. *IEEE Trans. Electron Devices* **2007**, *54*, 1825–1830. [[CrossRef](#)]
18. Dora, Y.; Chakraborty, A.; McCarthy, L.; Keller, S.; DenBaars, S.; Mishra, U. High breakdown voltage achieved on AlGaIn/GaN HEMTs with integrated slant field plates. *IEEE Electron Device Lett.* **2006**, *27*, 713–715. [[CrossRef](#)]
19. Huo, W.; Liang, H.; Lu, Y.; Han, Z.; Zhu, R.; Sui, Y.; Wang, T.; Mei, Z. Dual-active-layer InGaZnO high-voltage thin-film transistors. *Semicond. Sci. Technol.* **2021**, *36*, 065021. [[CrossRef](#)]
20. Yang, G.; Li, M.; Yu, Z.; Xu, Y.; Sun, H.; Liu, S.; Sun, W.; Wu, W. High-Voltage a-IGZO TFTs With the Stair Gate-Dielectric Structure. *IEEE Trans. Electron Devices* **2021**, *68*, 4462–4466. [[CrossRef](#)]
21. Karim, K.S.; Servati, P.; Nathan, A. High voltage amorphous silicon TFT for use in large area applications. *Microelectron. J.* **2004**, *35*, 311–315. [[CrossRef](#)]
22. Martin, R.A.; Da Costa, V.M.; Hack, M.; Shaw, J.G. High-voltage amorphous silicon thin-film transistors. *IEEE Trans. Electron Devices* **1993**, *40*, 634–644. [[CrossRef](#)]
23. Unagami, T.; Kogure, O. High-voltage TFT fabricated in recrystallized polycrystalline silicon. *IEEE Trans. Electron Devices* **1988**, *35*, 314–319. [[CrossRef](#)]
24. Yu, M.-J.; Lin, R.-P.; Chang, Y.-H.; Hou, T.-H. High-Voltage Amorphous InGaZnO TFT With Al<sub>2</sub>O<sub>3</sub> High-k Dielectric for Low-Temperature Monolithic 3-D Integration. *IEEE Trans. Electron Devices* **2016**, *63*, 3944–3949. [[CrossRef](#)]
25. Chow, E.M.; Lu, J.P.; Ho, J.; Shih, C.; De Bruyker, D.; Rosa, M.; Peeters, E. High voltage thin film transistors integrated with MEMS. *Sens. Actuators A* **2006**, *130*, 297–301. [[CrossRef](#)]
26. Li, X.; Liu, C.; Liu, C.; Ou, H.; She, J.; Deng, S.; Chen, J. Kilo-voltage thin-film transistors for driving nanowire field emitters. *IEEE Electron Device Lett.* **2020**, *41*, 405–408. [[CrossRef](#)]
27. Li, X.; Liu, C.; Kong, J.; Ou, H.; She, J.; Deng, S.; Chen, J. Widely Adjusting the Breakdown Voltages of Kilo-voltage Thin Film Transistors. *IEEE Electron Device Lett.* **2022**, *43*, 240–243. [[CrossRef](#)]
28. Hong, W.-C.; Ku, C.-J.; Li, R.; Abbaslou, S.; Reyes, P.; Wang, S.-Y.; Li, G.; Lu, M.; Sheng, K.; Lu, Y. MgZnO high voltage thin film transistors on glass for inverters in building integrated photovoltaics. *Sci. Rep.* **2016**, *6*, 34169. [[CrossRef](#)]

29. Hong, W.-C.; Zhang, Y.; Wang, S.-Y.; Li, Y.; Alim, N.; Du, X.; Mei, Z.; Lu, Y. ZnO flexible high voltage thin film transistors for power management in wearable electronics. *J. Vac. Sci. Technol. B Nanotechnol. Microelectron. Mater. Process. Meas. Phenom.* **2018**, *36*, 050601. [[CrossRef](#)]
30. Mativenga, M.; Jun, H.; Choe, Y.; Um, J.G.; Jang, J. Circular structure for high mechanical bending stability of a-IGZO TFTs. *IEEE J. Electron Devices Soc.* **2017**, *5*, 453–457. [[CrossRef](#)]
31. Huo, W.; Mei, Z.; Sui, Y.; Han, Z.; Wang, T.; Liang, H.; Du, X. Flexible transparent InGaZnO thin-film transistors on muscovite mica. *IEEE Trans. Electron Devices* **2019**, *66*, 2198–2201. [[CrossRef](#)]
32. Mativenga, M.; Ha, S.H.; Geng, D.; Kang, D.H.; Mruthyunjaya, R.K.; Heiler, G.N.; Tredwell, T.J.; Jang, J. Infinite output resistance of Corbino thin-film transistors with an amorphous-InGaZnO active layer for large-area AMOLED displays. *IEEE Trans. Electron Devices* **2014**, *61*, 3199–3205. [[CrossRef](#)]
33. Joo, H.-J.; Shin, M.-G.; Kwon, S.-H.; Jeong, H.-Y.; Jeong, H.-S.; Kim, D.-H.; Jin, X.; Song, S.-H.; Kwon, H.-I. High-gain complementary inverter based on Corbino p-type tin monoxide and n-type indium-gallium-zinc oxide thin-film transistors. *IEEE Electron Device Lett.* **2019**, *40*, 1642–1645. [[CrossRef](#)]
34. Geng, R.; Gong, Y. High performance active image sensor pixel design with circular structure oxide TFT. *J. Semicond.* **2019**, *40*, 022402. [[CrossRef](#)]
35. Deegan, R.D.; Bakajin, O.; Dupont, T.F.; Huber, G.; Nagel, S.R.; Witten, T.A. Capillary flow as the cause of ring stains from dried liquid drops. *Nature* **1997**, *389*, 827–829. [[CrossRef](#)]
36. Zhang, L.; Liu, H.; Zhao, Y.; Sun, X.; Wen, Y.; Guo, Y.; Gao, X.; Di, C.a.; Yu, G.; Liu, Y. Inkjet printing high-resolution, large-area graphene patterns by coffee-ring lithography. *Adv. Mater.* **2012**, *24*, 436–440. [[CrossRef](#)]
37. Liu, C.; Li, X.; Luo, Y.; Wang, Y.; Hu, S.; Liu, C.; Liang, X.; Zhou, H.; Chen, J.; She, J. How Materials and Device Factors Determine the Performance: A Unified Solution for Transistors with Nontrivial Gates and Transistor–Diode Hybrid Integration. *Adv. Sci.* **2021**, *9*, 2104896. [[CrossRef](#)]
38. Qiang, L.; Yao, R. A new definition of the threshold voltage for amorphous InGaZnO thin-film transistors. *IEEE Trans. Electron Devices* **2014**, *61*, 2394–2397. [[CrossRef](#)]
39. Byun, Y.H.; Den Boer, W.; Yang, M.; Gu, T. An amorphous silicon TFT with annular-shaped channel and reduced gate-source capacitance. *IEEE Trans. Electron Devices* **1996**, *43*, 839–841. [[CrossRef](#)]
40. Neamen, D.A. *Semiconductor Physics and Devices: Basic Principles*, 4th ed.; McGraw-Hill: New York, NY, USA, 2003; pp. 446–447.

**Disclaimer/Publisher’s Note:** The statements, opinions and data contained in all publications are solely those of the individual author(s) and contributor(s) and not of MDPI and/or the editor(s). MDPI and/or the editor(s) disclaim responsibility for any injury to people or property resulting from any ideas, methods, instructions or products referred to in the content.